

NC-Cap/PSR[™] (Primary Side Regulation) CV/CC Controller

FEATURES

- Proprietary "Audio Noise Cancellation" Control
- Built-in AC Line CC Compensation
- Proprietary "Zero-Output Startup" Control
- Proprietary "Smart Output Short Protection"
- ◆ 11.7V UVLO Hysteresis Window
- ♦ FB Short Protection
- Direct Drive of Low Cost BJT
- ◆ Proprietary NC-Cap/PSR[™] Control
- Proprietary Cable Drop Compensation
- Multi Mode Control
- Wide VDD Operating Range (7V to 24V)
- \pm 5% CC and CV Precision
- Low Standby Power Under 70mW
- Compensate for Transformer Inductance Tolerances and Line Voltage Variation
- Cycle-by-Cycle Current Limiting
- Leading Edge Blanking (LEB)
- Built-in Soft Start
- Pin Floating Protection
- VDD UVLO, OVP & Clamp

APPLICATIONS

- Battery chargers
- Replaces linear transformer and RCC SMPS
- Small power adapter
- ♦ AC/DC LED lighting

GENERAL DESCRIPTION

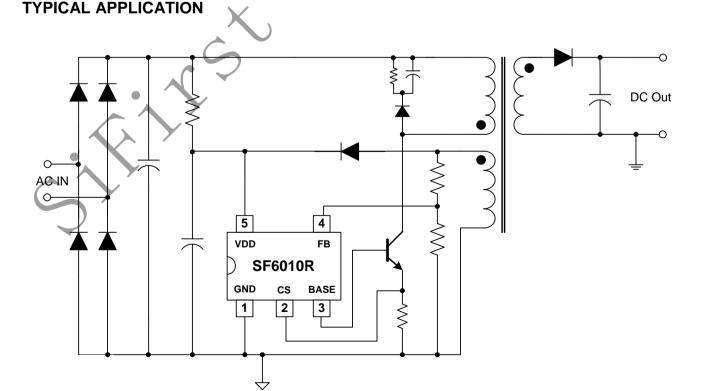
SF6010R is a high precision, highly integrated DCM (Discontinuous Conduction Mode) Primary Side Regulation (PSR) controller for offline small power converter applications. It can directly drive power BJT, which can further lower system cost. The IC integrates proprietary "Audio Noise Cancellation" control for audio noise free operation.

SF6010R has built-in proprietary **NC-Cap/PSR[™]** control for CV control, which eliminates external compensation or filtering capacitor. It has built-in cable drop compensation function, which can provide excellent CV performance. The IC uses Multi Mode Control to improve efficiency and reliability and to decrease audio noise energy @ light loadings. The IC can achieve less than 70mW standby power.

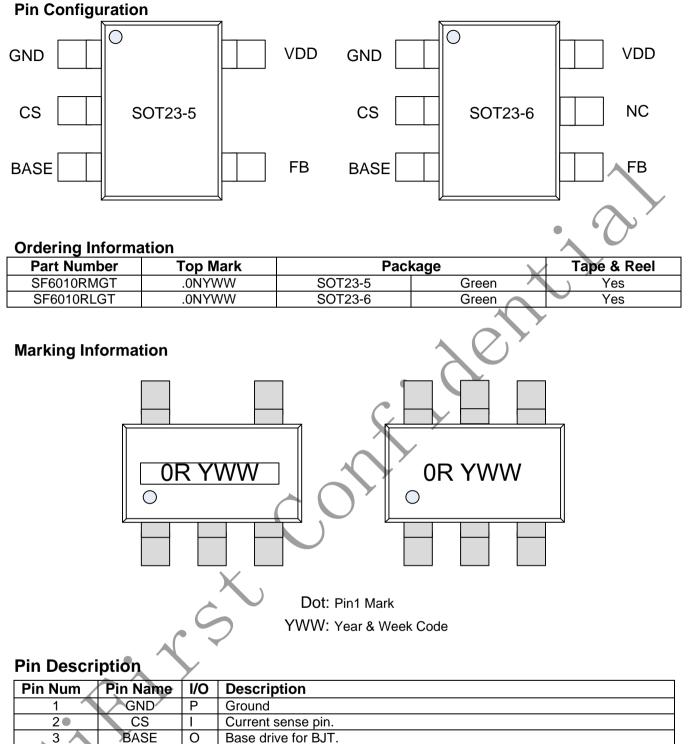
SF6010R integrates proprietary "Zero-Output Startup" control to achieve startup when output is near zero voltage. The IC also has proprietary "Smart Output Short Protection", which can limit maximum power when output is short circuit.

SF6010R integrates functions and protections of FB Short Protection, Under Voltage Lockout (UVLO), VDD Over Voltage Protection (VDD OVP), Soft Start, Cycle-by-cycle Current Limiting (OCP), Pin Floating Protection, VDD Clamping.

SF6010R is available in SOT23-5 and SOT23-6 packages.





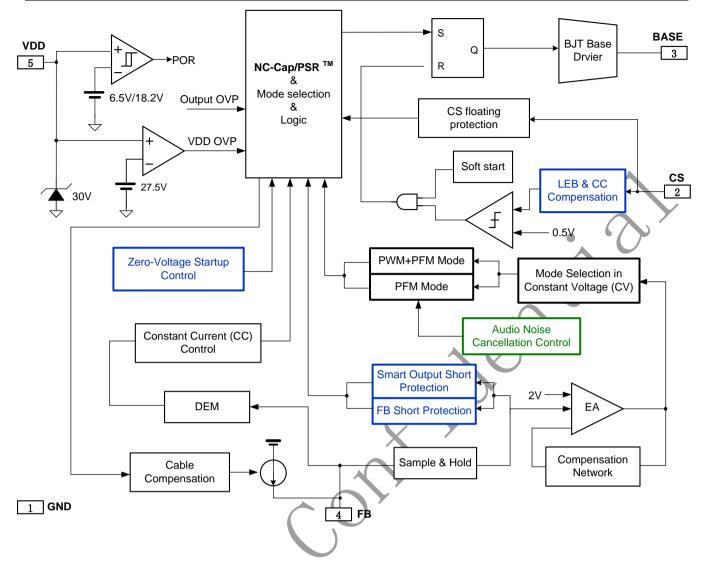


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4	FB	1	System feedback pin. This control input regulates both the output voltage
	r		in CV mode and output current in CC mode based on the flyback voltage
			of the auxiliary winding.
5	VDD	Р	IC power supply pin.

Block Diagram



SF6010R



Absolute Maximum Ratings (Note 1)

Parameter	Value	Unit
VDD DC Supply Voltage	30	V
VDD DC Clamp Current	10	mA
CS, BASE voltage range	-0.3 to 7	V
FB voltage range	-0.7 to 7	V
Package Thermal Resistance (SOT-23-5)	300	°C/W
Package Thermal Resistance (SOT-26)	250	°C/W
Maximum Junction Temperature	150	°C
Operating Temperature Range	-40 to 85	О°
Storage Temperature Range	-65 to 150	О°
Lead Temperature (Soldering, 10sec.)	260	О°
ESD Capability, HBM (Human Body Model)	3	kV
ESD Capability, MM (Machine Model)	250	V

Recommended Operation Conditions (Note 2)

Parameter	Value	Unit
Supply Voltage, VDD	7 to 24	V
Operating Ambient Temperature	-40 to 85	°C
Maximum Switching Frequency	70K	Hz

ELECTRICAL CHARACTERISTICS

 $(T_A = 25^{\circ}C, VDD=20V, if not otherwise noted)$



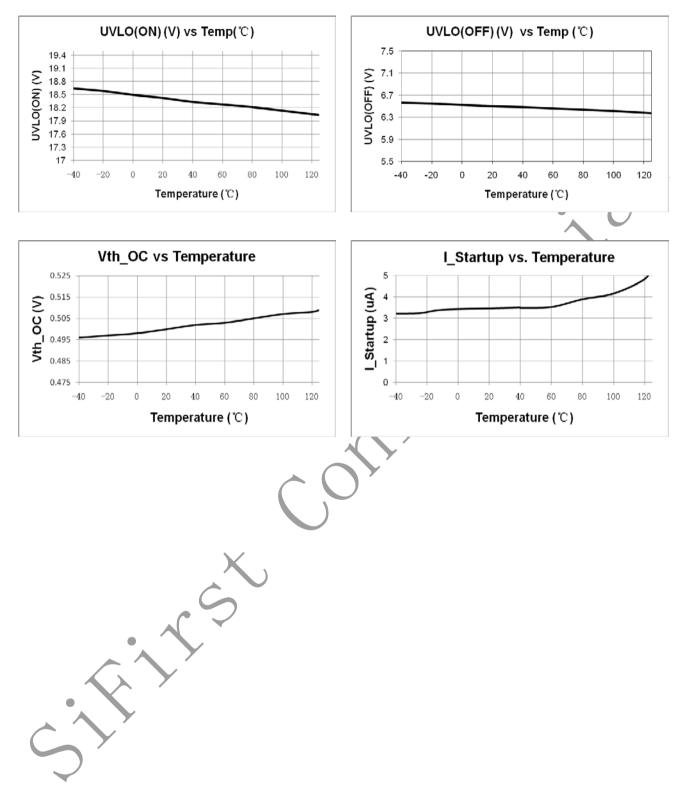
				-		
Symbol	Parameter	Test Conditions	Min	Тур	Max	Unit
Supply Voltage						
I_Startup	VDD Start up Current	VDD=UVLO(ON)-1V,		3	20	uA
		Measure current into VDD				
I_VDD_Op	Operation Current	VDD=20V		0.8	1.5	mA
UVLO(ON)	VDD Under Voltage		17	18.2	19.5	V
	Lockout Exit (Startup)					
UVLO(OFF)	VDD Under Voltage		5.9	6.5	7	V
	Lockout Enter					
VDD_OVP	VDD Over Voltage		25	27.5	30	V
	Protection trigger					
V _{DD} _Clamp	VDD Zener Clamp	$I(V_{DD}) = 10 \text{ mA}$	28	30	32	V
	Voltage					
T_Softstart	Soft Start Time			2		mSec
Feedback Input					•	\mathcal{O}
V _{FB} _EA_Ref	Internal Error		1.98	2.0	2.02	V
	Amplifier(EA)					
	reference input					
V _{FB} _DEM	Demagnetization			25		mV
	comparator threshold					
T _{min} _OFF	Minimum OFF time			2		uSec
		A)		
T _{max} _OFF	Maximum OFF time			1.2		mSec
V _{FB} _Short	Output Short Circuit			1.2		V
	Threshold					
T _{FB} _Short	Output Short Circuit	C		13		mSec
	Debounce Time					
T_{CC}/T_{DEM}	Ratio between			2		
	switching period in	$\sim \gamma$				
	CC mode and					
	demagnetization time					
I _{Cable} _max	Max Cable			60		uA
	compensation current					
Current Sense I	nput Section (CS Pin)					
T_blanking	CS Input Leading			500		nSec
-	Edge Blanking Time					
Vth_OC_max	Max. Current limiting		490	500	510	mV
	threshold					
T _D OC	Over Current			100		nSec
	Detection and Control					
	Delay					
Base Drive Sect		1	1			
				50		m ^
I _{BASE} _max	Max. Base Sourcing			50		mA
	Current		0.5	4	4.5	A
I _{BASE} pre_OFF	Base Sourcing		0.5	1	1.5	mA
	Current after Pre-off			4		
Rdson_low	Output Low Level ON			1		Ω
	Resistance					

Note 1. Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

Note 2. The device is not guaranteed to function outside its operating conditions.



CHARACTERIZATION PLOTS



- 5 -



OPERATION DESCRIPTION

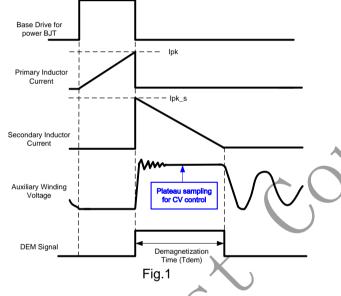
SF6010R is a high performance, highly integrated DCM (Discontinuous Conduction Mode) Primary Side Regulation (PSR) controller. The built-in high precision CV/CC control makes it very suitable for offline small power converter applications.

PSR Technology Introduction

Assuming the system works in DCM mode, the power transfer function is given by

$$P = \frac{\eta}{2} \times L_m \times I_{pk}^2 \times f_s = V_o \times I_o \quad \text{(Eq.1)}$$

In the equation above, P is output power, Vo and Io are system output voltage and current respectively, η is system power transfer efficiency, Lm is transformer primary inductance, fs is system switching frequency, Ipk is primary peak current in a switching cycle. The following figure illustrates the waveform in a switching cycle.



In the figure shown above, the IC generates a demagnetization signal (DEM) in each switching cycle through auxiliary winding. Tdem is demagnetization time for CV/CC control. In DCM mode, Tdem can be expressed as;

$$\frac{V_o}{L_m} \times T_{dem} = \frac{N_s}{N_p} \times I_{pk}$$
(Eq.2)

In Eq.2, Np and Ns are primary and secondary winding turns respectively.

Combined with Eq.1 and Eq. 2, the average output current can be expressed as:

$$I_o = \frac{\eta}{2} \times I_{pk} \times \frac{N_P}{N_S} \times f_S \times T_{dem}$$
 (Eq.3)

CC (Constant Current) Control Scheme

From Eq.3, it can be easily seen that there are two ways to implement CC control: one is PFM (Pulse Frequency Modulation), the control scheme is to keep lpk to be constant, let the product of Ts and Tdem (fs*Tdem) to be a constant. In this way, lo will be a value independent to the variation of Vo, Lm, and line input voltage. Another realization method is PWM duty control, the control scheme is to keep fs to be constant, let the product of Tdem and lpk (Tdem*lpk) to be a constant, in another words, by modulating system duty cycle to realize a constant lo independent to the variation of Vo, Lm and line voltages.

SF6010R adopts PFM for CC control, the product of Ts and Tdem is given by

$$f_s \times T_{dem} = 0.5$$
 (Eq.4)

CV (Constant Voltage) Control Scheme

CV control should sample the plateau of auxiliary winding voltage in flyback phase, as shown in Fig.1 The CV control has many implementations, for example, PWM, or PFM, or a combination of both one. In SF6010R, the CV control adopts proprietary multi mode control, as mention below.

Startup Current / Startup Control / Operating Current

Startup current of SF6010R is designed to be very low (typically 3uA) so that VDD could be charged up above UVLO(ON) threshold level and device starts up quickly. The operating current in SF6010R is as small as 0.8mA (typical). The small operating current results in higher efficiency and reduces the VDD hold-up capacitance requirement.

◆ NC-Cap/PSR-II[™] Introduction

• \pm 4% Precision CV/CC Performance

SF6010R can achieve less than \pm 4% variation of CC/CV precision due to the built-in CV accuracy improvement and CC line and load compensation, as shown in Fig.2.

• Proprietary "Audio Noise Cancellation" Control

SF6010R has a proprietary "Audio Noise Cancellation" control, which can achieve audio noise free operation in the whole loading range.

• Built-in Fast Dynamic Response Control to Meet USB Charge Requirements

In SF6010R, a fast dynamic response control is integrated to improve system dynamic response performance, thus the charger system can meet the USB charge requirements.

Smart Output Short Protection

The output short circuit protection of conventional PSR system is based on the coupling between auxiliary winding and secondary winding. When output is short, the auxiliary winding cannot provide enough energy to the IC any more. In this way, the system will enter into auto-recovery mode protection. However, the IC may be wrongly supplied if the leakage inductance of the primary



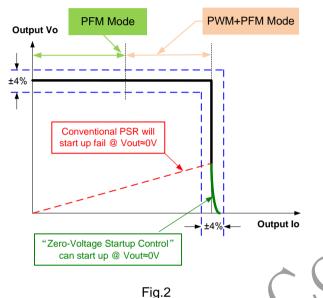
winding is large enough.

In SF6010R, if output short circuit occurs, the IC will detect the situation and enter into auto-recovery mode protection.

Proprietary Zero-Output Startup Control

Conventional PSR system may suffer startup failure when output voltage is near zero voltage, which means that there is a gap between OCP (CC point in PSR CV/CC system) and full loading. Larger OCP gap causes larger system cost.

In SF6010R, a proprietary "Zero-Output Startup Control" is adopted to achieve successful startup @ Vout≈0V, as shown in Fig.2.



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No External Compensation/Filtering Capacitor Needed

SF6010R uses a proprietary control to eliminate external compensation capacitor, which can simplify system design and lower system cost.

• Maximum 50V Output for LED Lighting

SF6010R can support maximum 50V output, which can be used in AC/DC LED lighting.

Proprietary Cable Drop Compensation

SF6010R has a proprietary built-in cable voltage drop compensation block which can provide a constant output voltage at the end of the cable over the entire load range in CV mode.

Multi Mode PSR Control for High Reliability, High Efficiency

Conventional pure PFM controlled PSR system may suffer transformer saturation issue when heavy loading. In SF6010R, a proprietary multi mode control is adopted to suppress this issue, as shown in Fig.2. Around the full load, the system operates in PWM+PFM mode, which improve the system reliability. Under normal to light load conditions, the IC operates in PFM mode to achieve excellent regulation and high efficiency.

Soft Start

SF6010R features an internal 2ms (typical) soft start that slowly increases the threshold of cycle-bycycle current limiting comparator during startup sequence. Every startup process is followed by a soft start activation.

Leading Edge Blanking (LEB)

Each time the power BJT is switched on, a turn-on spike occurs across the sensing resistor. To avoid premature termination of the switching pulse, an internal leading edge blanking circuit is built in. During this blanking period (500ns, typical), the cycle-by-cycle current limiting comparator is disabled and cannot switch off the base driver.

Minimum and Maximum OFF Time

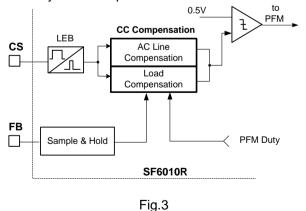
In SF6010R, a minimum OFF time (typically 2us) is implemented to suppress ringing when BASE drive is pull off. The maximum OFF time in SF6010R is typically 1.2ms, which provides a large range for frequency reduction. In this way, low standby power can be achieved.

Pin Floating Protection

In SF6010R, if pin floating situation occurs, the IC is designed to have no damage to system.

Built-in Load and AC Line CC Compensation

In conventional PSR system, the output CC (Constant Current) point can vary with output and AC line voltage. In SF6010R the IC has built-in blocks to compensate the variation, as shown in Fig3. The IC can adjust CC point based on sensed output voltage and PFM duty. In this way, CC accuracy can be improved.

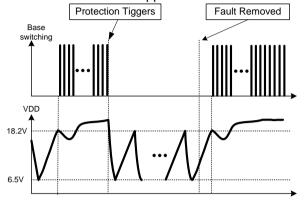


Auto Recovery Mode Protection

As shown in Fig.4, once a fault condition is detected, switching will stop. This will cause VDD to fall because no power is delivered form the auxiliary winding. When VDD falls to UVLO(off) (typical 6.5V), the protection is reset and the



operating current reduces to the startup current, which causes VDD to rise, as shown in Fig.4. However, if the fault still exists, the system will experience the above mentioned process. If the fault has gone, the system resumes normal operation. In this manner, the auto restart can alternatively enable and disable the switching until the fault condition is disappeared.





VDD OVP(Over Voltage Protection)

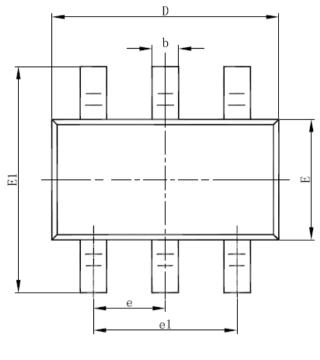
VDD OVP is implemented in SF6010R and it is a protection of auto-recovery mode.

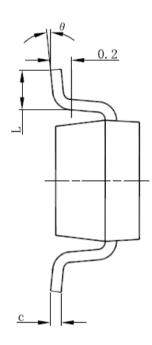
♦ Base Drive

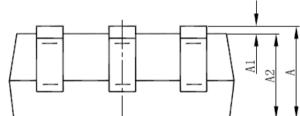
SF6010R can directly drive external power BJT. The maximum base drive current is typical 50mA. The base drive current is dynamic controlled to optimize performance.

PACKAGE MECHANICAL DATA

SOT-23-6L PACKAGE OUTLINE DIMENSIONS





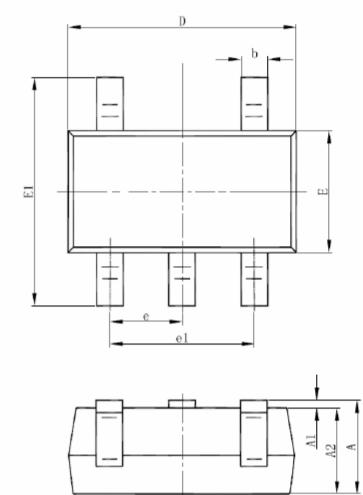


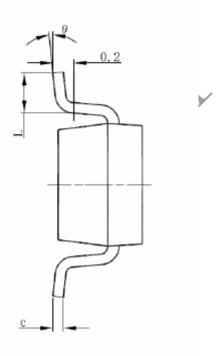
)			
Symbol	Dimensions	n Millimeters	Dimensions In Inches		
Symbol	Min	Max	Min	Max	
A	1.000	1.300	0.039	0.051	
A1	0.000	0.150	0.000	0.006	
A2	1.000	1.200	0.039	0.047	
b	0.300	0.500	0.012	0.020	
C	0.100	0.200	0.004	0.008	
D	2.800	3.020	0.110	0.119	
EY	1.500	1.700	0.059	0.067	
E1	2.600	3.000	0.102	0.118	
e	0.950	(BSC)	0.037 (BSC)		
e1	1.800	2.000	0.071	0.079	
L	0.300	0.600	0.012	0.024	
θ	0°	8°	0°	8º	

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PACKAGE MECHANICAL DATA

SOT-23-5L PACKAGE OUTLINE DIMENSIONS





Symbol	Dimensions I	In Millimeters	Dimensions In Inches				
Symbol	Min 🗸	Max	Min	Max			
A	1.050	1.250	0.041	0.049			
A1	0.000	0.100	0.000	0.004			
A2	1.050	1.150	0.041	0.045			
b	0.300	0.500	0.012	0.020			
C	0.100	0.200	0.004	0.008			
	2.820	3.020	0.111	0.119			
E	1.500	1.700	0.059	0.067			
E1	2.650	2.950	0.104	0.116			
е	0.950	(BSC)	0.037 (BSC)				
e1	1.800	2.000	0.071	0.079			
L	0.300	0.600	0.012	0.024			
θ	0°	8°	0°	8°			

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